

Description

VOLTAGE CLAMPER CAPABLE OF CONTROLLING A VOLTAGE DROP ACCORDING TO AN EXTERNAL INPUT VOLTAGE

BACKGROUND OF INVENTION

[0001] 1. Field of the Invention

[0002] The present invention provides a voltage clamper, and more particularly, to a voltage clamper capable of providing a corresponding voltage drop according to an external input voltage.

[0003] 2. Description of the Prior Art

[0004] With the progressive development of semiconductor processes, many different circuits are integrated into integrated circuits to drive the development of electronic products. For example, one memory chip may comprise a plurality of memory cells for storing data. Owing to the development of semiconductor processes, more memory

cells are accommodated in a same area on a memory chip. Typically, the operation voltages of internal devices are limited in a voltage range according to the spec of integrated circuits. For example, the operation voltage of the above-mentioned memory chip must be limited in a voltage range to allow the memory chip to function normally. When the operation voltage supplied to the memory chip is too high, structural damage to the memory cells in the memory chip may cause reliability issues in data storage of memory cells. Oppositely, the operation voltage may not be able to successfully drive the memory cells to store data in a predetermined period of time when the operation voltage supplied to the memory chip is too low. Therefore, the memory chip must operate under a low clock. In other words, an operation voltage that is too low will affect the performance of the memory chip greatly.

[0005] Generally speaking, the same memory chip can be applied to different devices and used for storing data temporarily. However, different devices may be supplied with different external voltages. For example, the power supply module of one device provides a voltage level of 3.6V, but the power supply module of another device provides a voltage level of 1.6V. Therefore, the prior art memory chip utilizes

a voltage drop circuit to transform the external voltage to the internal operation voltage that is applicable to the memory chip. For example, the voltage drop circuit may generate a fixed voltage drop of 1V. Under the circumstances, the range of the operation voltage of the memory chip to function normally is 2.6V–1.6V, based on the spec of the voltage drop circuit. In other words, the memory chip having such a voltage drop circuit can be only applied to devices supplied with an external voltage ranging from 3.6V to 2.6V. When the memory chip having such a voltage drop circuit is applied to a device supplied with an external voltage of 4V, the operation voltage of the memory chip will exceed the normally functional range of the operation voltage of the memory chip (2.6V–1.6V), since the operation voltage of the memory chip, used for driving the internal memory cells, is 3V after the voltage drop circuit applies a voltage drop of 1V to the external voltage. As a result, reliability issues arise when the memory chip is storing data. Similarly, the operation voltage of the memory chip, used for driving the internal memory cells, is 1V after the voltage drop circuit applies a voltage drop of 1V to the external voltage when the memory chip having such a voltage drop circuit is applied to a device sup-

plied with an external voltage of 2V. Since 1V is not within the normally functional range of the operation voltage of the memory chip (2.6V–1.6V), the performance of the memory chip is greatly affected due to this operation voltage that is too low, as mentioned previously.

[0006] Since a fixed voltage drop is generated by the voltage drop circuit utilized in the prior art memory chip, the application range of the memory chip is limited by the fixed voltage drop. As mentioned previously, the memory chip can be only applied to devices supplied with an external voltage ranging from 3.6V to 2.6V because a 1V voltage drop is generated by the voltage drop circuit and the normally functional range of the operation voltage of the memory chip is 2.6V–1.6V. When the memory chip is applied to a device supplied with an external voltage of 4V, the voltage drop circuit on the memory chip needs to be re-designed to lift the voltage drop generated by itself. Similarly, when the memory chip is applied to a device supplied with an external voltage of 2V, the voltage drop circuit on the memory chip also needs to be re-designed to reduce the voltage drop generated by itself. Therefore, the production cost of the memory chip is greatly raised to make the memory chip not competitive.

SUMMARY OF INVENTION

[0007] It is therefore a primary objective of the present invention to provide a voltage clamper to determine a corresponding voltage drop according to an external input voltage to resolve the above-mentioned problems.

[0008] According to the claimed invention, a voltage clamper for generating an output voltage by adjusting an input voltage is disclosed. The voltage clamper comprises a bias circuit for generating at least a bias voltage according to the input voltage, a voltage drop circuit for applying a voltage drop to the input voltage, and a voltage detection circuit electrically connected to the voltage drop circuit and the bias circuit for generating the output voltage through adjusting the voltage drop generated from the voltage drop circuit according to the bias voltage.

[0009] According to the claimed invention, a voltage adjusting method for generating an output voltage by adjusting an input voltage is disclosed. The voltage adjusting method comprises setting a plurality of voltage segments corresponding to a plurality of different voltage drop setting values, and utilizing one of the voltage drop setting values to trigger a voltage difference between the output voltage and the input voltage corresponding to the voltage drop

setting value when the input voltage is within one of the voltage segments.

[0010] It is an advantage of the claimed invention that the present invention voltage clamper dynamically determines the voltage drop applied in the voltage drop operation according to the voltage level of the external input voltage, rather than applying a fixed voltage drop. Therefore, the present invention voltage clamper can maintain the output voltage within the corresponding range of the operation voltage of the device utilizing the present invention voltage clamper, no matter if the external input voltage has a high voltage level or a low voltage level. As a result, the phenomena of insufficient voltage drop and over high voltage drop, which usually occur when utilizing the prior art voltage clamper, do not occur when utilizing the present invention voltage clamper.

[0011] These and other objectives of the present invention will no doubt become obvious to those of ordinary skill in the art after having read the following detailed description of the preferred embodiment that is illustrated in the various figures and drawings.

BRIEF DESCRIPTION OF DRAWINGS

[0012] Fig.1 is a function block diagram of a voltage clamper of

the present invention.

[0013] Fig.2 is a circuit diagram of the voltage clamper shown in Fig.1.

[0014] Fig.3 is a schematic diagram of the output voltage of the voltage clamper shown in Fig.2.

[0015] Fig.4 is a circuit diagram of the adjusting module shown in Fig.3.

DETAILED DESCRIPTION

[0016] Please refer to Fig.1. Fig.1 is a function block diagram of a voltage clamper 10 of the present invention. The present invention voltage clamper 10 comprises a bias circuit 12, a voltage detection circuit 14, and a voltage drop circuit 16. The bias circuit 12 is used for generating a bias voltage according to an input voltage V_{in} . The voltage detection circuit 14 determines how much voltage drop the voltage drop circuit 16 applies to the input voltage V_{in} according to the bias voltage so as to generate an output voltage V_{out} . In this preferred embodiment, a plurality of bias units 18 are installed in the bias circuit 12, a plurality of voltage detection units 20 are installed in the voltage detection circuit 14, and a plurality of voltage drop units 22 and a predetermined voltage drop unit 23 are installed in the voltage drop circuit 16. Without altering the teach-

ings of the invention, only three bias units 18a, 18b, 18n, three voltage detection units 20a, 20b, 20n, and three voltage drop units 22a, 22b, 22n are shown in Fig.1 for simplicity. When the input voltage V_{in} is input to the bias circuit 12, the bias unit 18a will generate a bias voltage V_1 according to the input voltage V_{in} , the bias unit 18b will generate a bias voltage V_2 according to the input voltage V_{in} , and the bias unit 18n will generate a bias voltage V_n according to the input voltage V_{in} . The voltage levels of the bias voltages V_1, V_2, V_n are different from each other. Assume that the bias voltage V_1 is greater than the bias voltage V_2 , and the bias voltage V_2 is greater than the bias voltage V_n . Although the bias voltages V_1, V_2, V_n change when the magnitude of the input voltage changes, the relative magnitude relationship of the bias voltages V_1, V_2, V_n does not change. For example, if the bias voltages V_1, V_2, V_n are 2V, 1.8V, and 1.5V, respectively, when the input voltage V_{in} is 5V, the bias voltages V_1, V_2, V_n are reduced to 1.6V, 1.4V, and 1.2V, respectively, when the input voltage V_{in} is reduced to 4V. Therefore, the bias voltage V_1 is still greater than the bias voltage V_2 , and the bias voltage V_2 is still greater than the bias voltage V_n .

[0017] The voltage detection unit 20a receives the bias voltage V_1

to generate a control signal D_1 according to the bias voltage V_1 , the voltage detection unit 20b receives the bias voltage V_2 to generate a control signal D_2 according to the bias voltage V_2 , and the voltage detection unit 20n receives the bias voltage V_n to generate a control signal D_n according to the bias voltage V_n . In this preferred embodiment, each voltage detection unit 20a, 20b, 20n in the voltage detection circuit 14 is used for detecting a same predetermined voltage level. That means, each voltage detection unit 20a, 20b, 20n sieves the bias voltages V_1, V_2, V_n according to the predetermined voltage level and decides whether to output the control signal D_1, D_2, D_n to activate the voltage drop unit 22a, 22b, 22n or not. Each voltage drop unit 22a, 22b, 22n is used to apply a different voltage drop to the input voltage V_{in} to adjust the voltage level of the output voltage V_{out} . For example, the voltage drop unit 22a may make the input voltage V_{in} generate a voltage drop of dV_1 . That means, the output voltage V_{out} will be approximately equal to $V_{in} - dV_1$ when the voltage drop unit 22a is activated. Similarly, the voltage drop unit 22b may make the input voltage V_{in} generate a voltage drop of dV_2 . That means, the output voltage V_{out} will be approximately equal to $V_{in} - dV_2$ when the volt-

age drop unit 22b is activated. The voltage drop unit 22n may make the input voltage V_{in} generate a voltage drop of dV_n . That means, the output voltage V_{out} will be approximately equal to $V_{in} - dV_n$ when the voltage drop unit 22n is activated. Therefore, the voltage drops between the output voltage V_{out} and the input voltage V_{in} are controlled by the voltage drop units 22a, 22b, 22n. Furthermore, a predetermined voltage drop unit 23 is installed in the voltage drop circuit 16 to apply an initial voltage drop to the input voltage V_{in} to affect the output voltage V_{out} when the voltage clamper 10 is activated.

[0018] Please refer to Fig.2. Fig.2 is a circuit diagram of the voltage clamper 10 shown in Fig.1. In order to illustrate more conveniently, only two bias units 18a, 18b, two voltage detection units 20a, 20b, two voltage drop units 22a, 22b, and one predetermined voltage drop unit 23 are shown in the voltage clamper 10 shown in Fig.2. As shown in Fig.1, please note that the numbers of the bias units, the voltage detection units, and the voltage drop units are not limited in the present invention voltage clamper 10. In this preferred embodiment, a current I_1 flowing through the bias unit 18a is different from a current I_2 flowing through the bias unit 18b under the same input voltage V_{in} by apply-

ing different connection structures to transistors 24a, 24b, 24c, 24d and transistors 26a, 26b, 26c, 26d in the bias units 18a, 18b. Finally, the bias voltage V_1 is controlled to be greater than the bias voltage V_2 . In addition, the bias units 18a, 18b may utilize other kind of circuits (such as a voltage dividing circuit comprising only resistors) to achieve the objective that different bias voltages V_1, V_2 are generated under the same input voltage V_{in} . The bias voltage V_1 generated by the bias unit 18a is input to an input terminal A of the voltage detection unit 20a. The voltage detection unit 20a thus decides whether to conduct transistors 28a, 28b or not according to the bias voltage V_1 . If the bias voltage V_1 is greater than a predetermined voltage level, the transistor 28b is turned on to drive the control signal D_1 to approach a high logic level "1". Oppositely, the transistor 28a is turned on and the transistor 28b is not turned on to drive the control signal D_1 to approach a low logic level "0" if the bias voltage V_1 is smaller than the predetermined voltage level. In addition, the bias voltage V_2 generated by the bias unit 18b is input to an input terminal B of the voltage detection unit 20b. Similarly, the voltage detection unit 20b thus decides whether to turn on the transistors 30a, 30b or not accord-

ing to the bias voltage V_2 . If the bias voltage V_2 is greater than the same predetermined voltage level, the transistor 30b is turned on to drive the control signal D_2 to approach the low logic level "0". Oppositely, the transistor 30a is turned on and the transistor 30b is not turned on to drive the control signal D_2 to approach the high logic level "1" if the bias voltage V_2 is smaller than the same predetermined voltage level.

[0019] In the voltage detection unit 20a, the operation of inverters 32a, 32b, 32c is similar to that of a prior art Schmitt trigger, and an inverter 32d functions as a buffer. In addition, a substrate, a source, and a drain of a transistor 28f are connected to ground. Therefore, the transistor 28f functions as a capacitor module to stabilize the control signal D1. When the transistor 28b is turned on, the loop formed by the inverters 32b, 32c will maintain an input terminal of the inverter 32d at the low logic level "0", and the transistor 28e is turned on. When the transistor 28b is not turned on, the loop formed by the inverters 32b, 32c will maintain the input terminal of the inverter 32d at the high logic level "1", and the transistor 28e is not turned on. For the voltage detection unit 20b, the operation of inverters 34a, 34b, 34c is similar to that of the prior art

Schmitt trigger, and inverters 34d, 34e function as buffers. In addition, a substrate, a source, and a drain of a transistor 30f are connected to ground. Therefore, the transistor 30f functions as a capacitor module to stabilize the control signal D_2 . When the transistor 30b is turned on, the loop formed by the inverters 34b, 34c will maintain an input terminal of the inverter 34d at the low logic level "0", and a transistor 30e is turned on. When the transistor 30b is not turned on, the loop formed by the inverters 34b, 34c will maintain the input terminal of the inverter 34d at the high logic level "1", and the transistor 30e is not turned on.

[0020] The voltage drop unit 22a comprises a transistor 36, and the voltage drop unit 22b comprises a transistor 38. In this preferred embodiment, the transistor 36 is a P-type metal-oxide-semiconductor (PMOS) transistor, and the transistor 38 is an N-type metal-oxide-semiconductor (NMOS) transistor. As well known by those skilled in the art, a P-type metal-oxide-semiconductor transistor is a good switch device, and an N-type metal-oxide-semiconductor transistor is a bad switch device, when transferring a high logic level "1". In other words, the voltage level at a drain of the transistor 36 is approxi-

mately equal to that at a source of the transistor 36 (that means the input voltage V_{in}) when the transistor 36 is turned on. However, the voltage level at a drain of the transistor 38 is greater than that at a source of the transistor 38 when the transistor 38 is turned on. In other words, the voltage level at the source of the transistor 38 is approximately equal to $V_{in} - V_t$, rather than the input voltage V_{in} . It is worth noting that V_t is the threshold voltage corresponding to a channel of the transistor 38. In addition, the predetermined voltage drop unit 23 comprises two transistors 40a, 40b in this preferred embodiment, and transistors 40a, 40b are both N-type metal-oxide-semiconductor transistors. As shown in Fig.2, a drain of the transistor 40a is connected to a gate of the transistor 40a, and a drain of the transistor 40b is connected to a gate of the transistor 40b. Therefore, the transistors 40a, 40b are always turned on and operate within a saturation region. As mentioned previously, an N-type metal-oxide-semiconductor transistor is a bad switch device when transferring a high logic level "1". If the transistors 40a, 40b have the same threshold voltage V_t as the transistor 38, the voltage level at a source of the transistor 40b eventually approaches $V_{in} - 2V_t$. In Fig.2, a

transistor 42 utilized in the voltage clamper 10 functions as a capacitor module to stabilize the voltage level of the output voltage V_{out} . In this preferred embodiment, a gate and a drain of the transistor are electrically connected to the output voltage V_{out} , and a substrate and a source of the transistor 42 are connected to ground. It is very obvious that the transistor 42 will be kept in a conductive state when the voltage clamper 10 is operating. Therefore, the transistor 42 may be regarded as a resistor in parallel with a capacitor. In comparison with the transistors 28f, 30f, the transistor 42 has a greater RC time constant to maintain the output voltage V_{out} more stably.

[0021] Please refer to Fig.2 and Fig.3. Fig.3 is a schematic diagram of the output voltage of the voltage clamper 10 shown in Fig.2. In Fig.3, the horizontal axis represents the input voltage V_{in} , and the vertical axis represents the output voltage V_{out} . It is known from the above description that the bias voltage V_1 output by the bias unit 18a is greater than the bias voltage V_2 output by the bias unit 18b with the same input voltage V_{in} , and the voltage detection units 20a, 20b detect a predetermined voltage level to decide whether to activate the voltage drop units 22a, 22b or not. When the input voltage V_{in} is equal to the

voltage level of $(V_s)_2$, the bias voltage V_2 is equal to the predetermined voltage level. At this time, the bias voltage V_1 is greater than the predetermined voltage level since the bias voltage V_1 is greater than the bias voltage V_2 . In other words, the transistor 30b in the voltage detection unit 20b will keep conducting until the bias voltage V_2 starts to be smaller than the predetermined voltage level. The control signal D_2 is thus at the high logic level "1" to activate the corresponding voltage drop unit 22b. When the input voltage V_{in} is equal to the voltage level of $(V_s)_1$, the bias voltage V_1 is equal to the predetermined voltage level. At this time, the bias voltage V_2 is smaller than the predetermined voltage level since the bias voltage V_2 is smaller than the bias voltage V_1 . In other words, the transistor 28b in the voltage detection unit 20a will keep conducting until the bias voltage V_1 starts to be smaller than the predetermined voltage level. The control signal D_1 is thus at the low logic level "0" to activate the corresponding voltage drop unit 22a. It is worth noting that the transistor 30b in the voltage detection unit 20b will be kept in a non-conductive state since the bias voltage V_2 is smaller than the predetermined voltage. The control signal D_2 is thus kept at the high logic level "1". As a result, the corre-

sponding voltage drop unit 22b is kept in an active state.

[0022] As shown in Fig.3, an oblique line L_1 represents that the output voltage V_{out} is equal to the input voltage V_{in} . When the input voltage V_{in} is greater than the voltage level of $(V_s)_2$, the bias voltages V_1, V_2 corresponding to the input voltage V_{in} are both greater than the above-mentioned predetermined voltage level. At this time, only the predetermined voltage drop unit 23 will affect the output voltage V_{out} . That means, the voltage difference between the output voltage V_{out} and the input voltage V_{in} corresponds to the voltage difference applied by the transistors 40a, 40b($2 \cdot V_t$), when the input voltage V_{in} is greater than the voltage level of $(V_s)_2$, and the relationship between the output voltage V_{out} and the input voltage V_{in} is shown in segment S1. When the input voltage V_{in} is smaller than the voltage level of $(V_s)_2$ and is greater than the voltage level of $(V_s)_1$, the bias voltage V_2 corresponding to the input voltage V_{in} is smaller than the predetermined voltage level, and the bias voltage V_1 corresponding to the input voltage V_{in} is still greater than the predetermined voltage level, as mentioned previously. At this time, both the voltage drop unit 22b and the predetermined voltage drop unit 23 are activated. It is worth noting that the predeter-

mined voltage drop unit 23 will apply a voltage difference of $2 \cdot V_t$ to the input voltage V_{in} , and the voltage drop unit 22b will only apply a voltage difference of V_t to the input voltage V_{in} . Since the transistor 42 is used as a capacitor module, the voltage drop unit 22b will charge the transistor 42 and trigger the voltage difference between the output voltage V_{out} and the input voltage V_{in} corresponding to the voltage difference applied by the transistor 38(V_t), and the relationship between the output voltage V_{out} and the input voltage V_{in} is shown in segment S2. When the input voltage V_{in} is smaller than the voltage level of (V_{s1}), the bias voltages V_1, V_2 corresponding to the input voltage V_{in} are both smaller than the predetermined voltage level, as mentioned previously. At this time, the voltage drop units 22a, 22b and the predetermined voltage drop unit 23 are all activated. It is worth noting that the predetermined voltage drop unit 23 will apply a voltage difference of $2 \cdot V_t$ to the input voltage V_{in} , and the voltage drop unit 22b will apply a voltage difference of V_t to the input voltage V_{in} , and the voltage drop unit 22a will not apply any voltage difference to the input voltage V_{in} . That means, the voltage drop unit 22a transfers the input voltage V_{in} to trigger the output voltage V_{out} . Since the tran-

sistor 42 is used as a capacitor module, the voltage drop unit 22a will charge the transistor 42 to trigger the output voltage V_{out} to be approximately equal to the input voltage V_{in} . The relationship between the output voltage V_{out} and the input voltage V_{in} is shown in segment S3.

[0023] It is worth noting that gates of the transistors 28c, 28d of the voltage detection unit 20a are all triggered by a control signal CEB in the voltage clamper 10 in Fig.2. Similarly, gates of the transistors 30c, 30d of the voltage detection unit 20b are all triggered by the same control signal CEB. The present invention voltage clamper 10 supports chip enable control to achieve the objective of low current consumption. The voltage clamper 10 can switch to a standby mode or a normal mode according to the external control signal CEB. For example, the voltage clamper 10 will enter the standby mode when the control signal CEB is at the high voltage level. At this time, the transistors 28c, 30c are not turned on, and the control signal CEB will turn on the transistors 28d, 30d. In other words, only the predetermined voltage drop unit 23 is activated when the voltage clamper 10 enters the standby mode, and the voltage drop units 22a, 22b cannot be turned on to adjust the output voltage V_{out} . Therefore, a greater

voltage difference (that is $2 * V_t$) exists between the output voltage V_{out} and the input voltage V_{in} . For a device utilizing the voltage clamper 10, the device will output the control signal CEB to the voltage clamper 10 when entering the standby mode. Since the output voltage V_{out} of the voltage clamper 10 in the standby mode is lower, the current consumed by the device in the standby mode is smaller to reduce power consumption. Oppositely, the control signal CEB will be at the low voltage level to trigger the voltage clamper 10 to enter the normal mode when the device wants to exit the standby mode and enter the normal mode. As shown in Fig.2, the transistors 28c, 30c are thus turned on to transfer the input voltage V_{in} to the transistors 28a, 30a. In addition, the transistors 28d, 30d will be kept in a non-conductive state. Therefore, the activation of the voltage drop units 22a, 22b are controlled by the bias voltages V_1, V_2 . That means, the relationship between the input voltage V_{in} and the output voltage V_{out} is shown as Fig.3.

[0024] If the present invention voltage clamper 10 is applied to a memory chip, and the normally functional range of the operation voltage of the memory chip is between the voltage level of V_{top} and the voltage level of V_{bot} , the memory

chip can operate smoothly when the input voltage V_{in} is between the voltage level of V_{top} and the voltage level of V_H ($V_H > V_{bot}$), as shown from the relationship between the output voltage V_{out} and the input voltage V_{in} in Fig.3.

Therefore, the greater the input voltage V_{in} is, the greater voltage drop between the input voltage V_{in} and the output voltage V_{out} is triggered by the voltage clamper 10. Oppositely, the smaller the input voltage V_{in} is, the smaller voltage drop between the input voltage V_{in} and the output voltage V_{out} is triggered by the voltage clamper 10. For example, the normally functional range of the operation voltage of the memory chip is 2.6V–1.6V. When the power supply module of a device supplies a high driving voltage of $(2.6+2*V_t)$, the voltage clamper 10 will help to transform the input voltage $(2.6+2*V_t)$ into an output voltage of 2.6V and transfer the output voltage (2.6V) to the memory chip to trigger the memory chip. The memory chip thus operates smoothly under such a high driving voltage. However, when the power supply module of a device supplies a low driving voltage of 1.6V, the voltage clamper 10 will not adjust the output voltage. That means, the output voltage is equal to the input voltage 1.6V, and the voltage clamper 10 will transfer the output voltage

(1.6V) to the memory chip to trigger the memory chip. As a result, the memory chip can function normally under a low external voltage.

[0025] When the driving voltage supplied by the power supply module in a device is between the voltage level of V_H and the voltage level of $(V_s)_2$, the memory chip utilizing the voltage clamber 10 can function normally in the device. Similarly, the memory chip utilizing the voltage clamber 10 can function normally in the device when the driving voltage supplied by the power supply module is between the voltage level of $(V_s)_1$ and the voltage level of $(V_s)_2$, and between the voltage level of V_{bot} and the voltage level of $(V_s)_1$. However, there is a problem when the driving voltage supplied by the power supply module approaches $(V_s)_1$ or $(V_s)_2$. It is known that the predetermined voltage level originally set by the voltage detection units 20a, 20b will control the voltage clamber 10 to trigger the output voltage V_{out} to generate changes of the voltage levels, when the voltage levels of the input voltage V_{in} are $(V_s)_1$ and $(V_s)_2$. In other words, the output voltage V_{out} will hop between two voltage levels if vibration of the driving voltage supplied by the power supply module occurs in the neighborhood of the voltage level of $(V_s)_1$ or $(V_s)_2$. As a result, the

memory chip generates unexpected errors. In order to resolve the problem, the voltage detection unit 20a further comprises an adjusting module 44 and the voltage detection unit 20b further comprises an adjusting module 46. The adjusting modules 44, 46 are used for adjusting the predetermined voltage levels detected by the voltage detection units 20a, 20b. Please refer to Fig.4. Fig.4 is a circuit diagram of the adjusting module shown in Fig.3. It is worth noting that only the adjusting module 44 is illustrated because the configuration and the operation of the adjusting module 44 and the adjusting module 46 are the same. The adjusting module 44 comprises a plurality of transistors 48. A drain of each of the transistors 48 is connected to a node A of the voltage detection unit 20a, and a gate of each of the transistor 48 is selectively connected to a source of the transistor or an input terminal A of the voltage detection unit 20a. When the gate of the transistor 48 is connected to the input terminal A of the voltage detection unit 20a, the transistor 48 is regarded as being in parallel with the transistor 28b. Therefore, the transistor 48 can be utilized to adjust the predetermined voltage level at the input terminal A detected by the voltage detection unit 20a. Oppositely, the transistor 48 can-

not be turned on and will not affect the operation of the voltage detection unit 20a when the gate of the transistor is connected to the source of the transistor. In this preferred embodiment, the gate of each of the transistor 48 is connected to the node A or the source of the transistor is programmed by an upper level metal layer. That means the metal layer is utilized to program the adjusting module 44. For example, the initial setting of the adjustment module 44 is achieved by programming the upper level metal layer through a mask pattern design during the semiconductor processes for forming the voltage clamper 10, and the initial setting of the adjusting module 44 is that the gates of half of the transistors 48 are connected to the input terminal A and the gates of half of the transistors 48 are connected to the sources of the corresponding transistors 48. At this time, the characteristic of the input voltage V_{in} and the output voltage V_{out} of the voltage clamper 10 is shown in Fig.3. If it is known that the driving voltage supplied by the power supply module in a device approaches the voltage level of $(V_s)_1$, another mask pattern design is utilized during the semiconductor processes for forming the voltage clamper 10. The numbers of the transistors 48 having the gates connected to

the sources of the transistors 48 and the numbers of the transistors 48 having the gates connected to the input terminal A are thus adjusted to bias the voltage level of $(V_s)_1$. In addition, the adjust module 44 can lower the voltage level of $(V_s)_1$ or lift the voltage level of $(V_s)_1$. Therefore, the problem that the output voltage V_{out} probably changes greatly due to the input voltage V_{in} approaching the original voltage level of $(V_s)_1$ is avoided. Since the operation of the adjusting module 46 is the same as that of the adjusting module 44, the adjust module 46 can lower the voltage level of $(V_s)_2$ or lift the voltage level of $(V_s)_2$ in this preferred embodiment. As a result, the problem that the output voltage V_{out} probably changes greatly due to the input voltage approaching the original voltage level of $(V_s)_2$ is avoided. In summary, the devices having the voltage clamper 10 can operate more stably by utilizing the adjusting modules 44, 46.

[0026] As mentioned previously, the operation of the voltage clamper 10 is to set the voltage detection units 20a, 20b, 20n to detect the same predetermined voltage level, and each of the bias unit 18a, 18b, 18n generates each of the different bias voltages V_1, V_2, V_n according to the input voltage V_{in} . Therefore, the magnitude of the input voltage

V_{in} is determined according to the bias voltages V_1, V_2, V_n and the predetermined voltage level to control the activation of the voltage drop units 22a, 22b, 22n. As a result, the voltage drop between the input voltage V_{in} and the output voltage V_{out} is adjusted. However, the objective of dynamically determining the voltage drop applied in the voltage drop operation according to the voltage level of the input voltage can be achieved by setting the voltage detection units 20a, 20b, 20n to detect different predetermined voltage levels and each of the bias units 18a, 18b, 18n to generate a same bias voltage according to the input voltage V_{in} . For example, each of the bias units 18a, 18b is set to generate a same bias voltage V_b according to the input voltage V_{in} . That means, the high input voltage V_{in} is transformed into the low bias voltage V_b . In addition, each of the voltage detection units 20a, 20b is set to detect different predetermined voltage levels of V_{d1}, V_{d2} , and the predetermined voltage level of V_{d1} is smaller than the predetermined voltage level of V_{d2} . It is very obvious that the greater the input voltage V_{in} is, the greater the bias voltage V_b is. Oppositely, the smaller the input voltage V_{in} is, the smaller the bias voltage V_b is. Therefore, the bias voltage V_b can be used to represent the magnitude of the

input voltage V_{in} . When the bias voltage V_b is greater than the predetermined voltage level of V_{d2} , only the predetermined voltage drop unit 23 is activated. When the bias voltage V_b is between the predetermined voltage level of V_{d1} and the predetermined voltage level of V_{d2} , both the predetermined voltage drop unit 23 and the voltage drop unit 22b are activated. When the bias voltage V_b is smaller than the predetermined voltage level of V_{d1} , the predetermined voltage drop unit 23 and the voltage drop units 22a, 22b are activated. Consequently, the above-mentioned relationship between the input voltage V_{in} and the output voltage V_{out} is shown in Fig.3. Therefore, the bias circuit 12 and the voltage detection circuit 14 may be set in such a manner as to trigger the voltage drop circuit 16, according to the voltage level of the input voltage V_{in} , so that the voltage difference between the output voltage V_{out} and the input voltage V_{in} corresponds to different voltage drops according to different voltage levels of the input voltage V_{in} .

[0027] Compared to the prior art voltage clamper, the present invention voltage clamper utilizes the bias circuit and the voltage detection circuit to judge the voltage level of the now applied external input voltage, and determine the

corresponding voltage difference between the output voltage and the input voltage according to the voltage level. According to the present invention voltage clamber, a plurality of voltage segments are set and each of the voltage segments corresponds to a specific voltage drop to adjust the output voltage. A greater voltage drop is applied to the input voltage corresponding to the voltage segment having a higher voltage level to generate the expected output voltage. Oppositely, a smaller voltage drop is applied to the input voltage corresponding to the voltage segment having a lower voltage level to generate the expected output voltage. In other words, the present invention voltage clamber will apply a greater voltage drop, according to the input voltage, to greatly reduce the output voltage when the input voltage has a high voltage level. Therefore, the problem that one device (such as a memory chip), triggered by the output voltage generated by the voltage clamber according to the external input voltage, cannot function normally owing to the output voltage exceeding the normally functional range of the operation voltage of the device is avoided. In addition, the present invention voltage clamber will not perform the voltage drop operation when the input voltage has a low

voltage level. Therefore, the performance of one device (such as a memory chip), triggered by the output voltage generated by the voltage clamper according to the external input voltage, is not greatly affected due to the output voltage being smaller than the normally functional range of the operation voltage of the device. In summary, the present invention voltage clamper dynamically determines the voltage drop applied in the voltage drop operation according to the voltage level of the external input voltage, rather than applying a fixed voltage drop. The present invention voltage clamper thus can maintain the output voltage within the range of the operation voltage of the device utilizing the present invention voltage clamper, no matter if the external input voltage has a high voltage level or a low voltage level. As a result, the phenomena of an insufficient voltage drop and an overly high voltage drop, which usually occur when utilizing the prior art voltage clamper, do not occur when utilizing the present invention voltage clamper.

[0028] Those skilled in the art will readily observe that numerous modifications and alterations of the device may be made while retaining the teachings of the invention. Accordingly, the above disclosure should be construed as limited

only by the metes and bounds of the appended claims.